

**EECS 240B**  
**Advanced Analog Integrated Circuits**  
**Lecture 3: Passives**

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Spring 2024

# Passive devices

## Summary

- **Resistors**
- **Capacitors**
- **Inductors**
- **Diodes**
- **BJT**

# Resistors

# Resistors types

- Silicide resistors

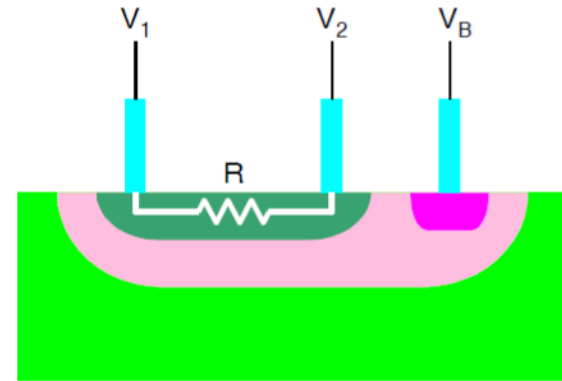
Layer	Sheet resistance
Aluminum	60 mΩ/□
Polysilicon	5 Ω/□
N+/P+ diffusion	5 Ω/□
N-well	1 kΩ/□

- Non-silicide (higher sheet resistance)

Layer	R/□ [Ω/□]	T <sub>C</sub> [ppm/°C] @ T = 25 °C	V <sub>C</sub> [ppm/V]	B <sub>C</sub> [ppm/V]
N+ poly	100	-800	50	50
P+ poly	180	200	50	50
N+ diffusion	50	1500	500	-500
P+ diffusion	100	1600	500	-500
N-well	1000	-1500	20,000	30,000

- Absolute resistance values varies by +-20%
  - Trimming needed if accuracy is of concern
- Temperature coefficient (TC, ohm/°C)
- Voltage coefficient (VC, ohm/V)
- Can mix different types to create zero-TC resistor

## P+ diffusion resistor



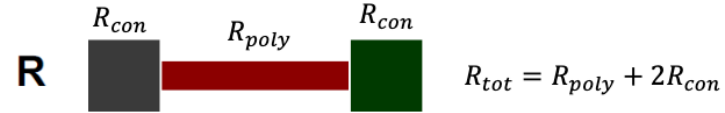
# Silicide

- Ion implantation to lower the resistance of source/drain and polysilicon material.
- Titanium silicide ( $\text{TiSi}_2$ ) is widely used silicide (self-aligned silicide), low resistivity (13-17  $\Omega\text{-cm}$ ) and high melting temperature (1540°C).
- Another widely used is  $\text{CoSi}_2$  and  $\text{PtSi}$ .

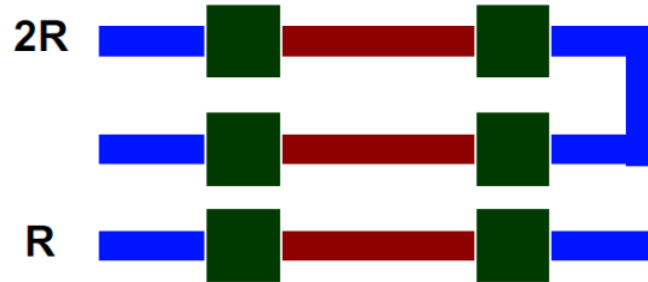
# Resistors

## Variations and mismatches

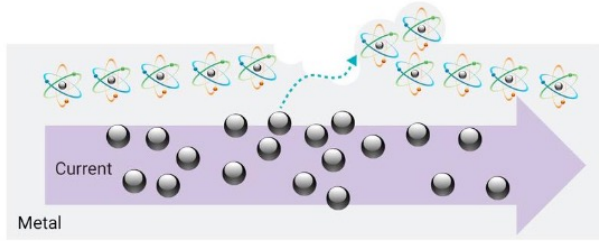
- Systematic, e.g. need a R and 2R.
- Batch variation
- Mismatches
  - Current direction matters!



VS

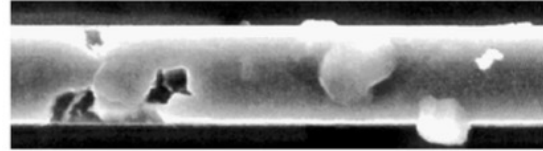


# Electron migration



*Courtesy of Synopsis*

*Need to make sure wire width is enough to handle electron-migration current density rules*



Resistance changes over time  
as a function of current density ( $\text{mA}/\mu\text{m}$ ),  
DC or AC (r.m.s), temperature

Check current density requirement (in DRM) offered by foundry,  
and put those equations into an excel sheet for calculation.

- Electron migration is related to current density flow within a metal.
- If the current density is high enough, the heat dissipated within the material will repeatedly break atoms from the structure and move them. This will create both 'vacancies' and 'deposits'. The vacancies can grow and eventually break circuit connections resulting in open- circuits, while the deposits can grow and eventually close circuit connections resulting in short-circuit.

Worst case...

Deposition of atoms  
(Hillocks) result in  
a short

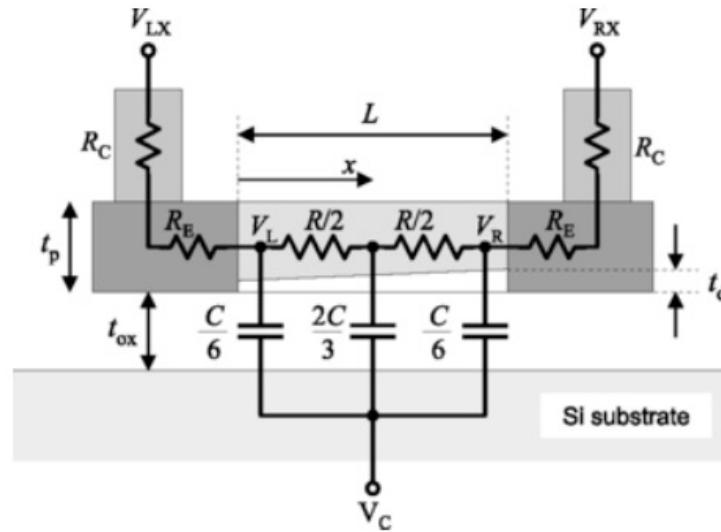


Depletion of atoms  
(Voids) result in  
an open



# Resistors

## Model



- Polysilicon resistors can have depletion due to electrostatic, causing resistance variation.
- My personal preference is “p+ poly”, to have same type as the silicon substrate.
- Make sure to include parasitic capacitance in your design (degrade bandwidth), included in the post-layout extraction.
- Use width  $> 2 \mu\text{m}$  if absolute resistance is of concern. Experience 20% drop even with  $w = 10 \mu\text{m}$ ...
- Make sure to include dummy resistor on each side to avoid under-cut during the lithography.
- Resistors can have “flicker ( $1/f$ )” noise, this is not always included in the model (some in RF model, but accuracy is questionable...).



# Capacitors

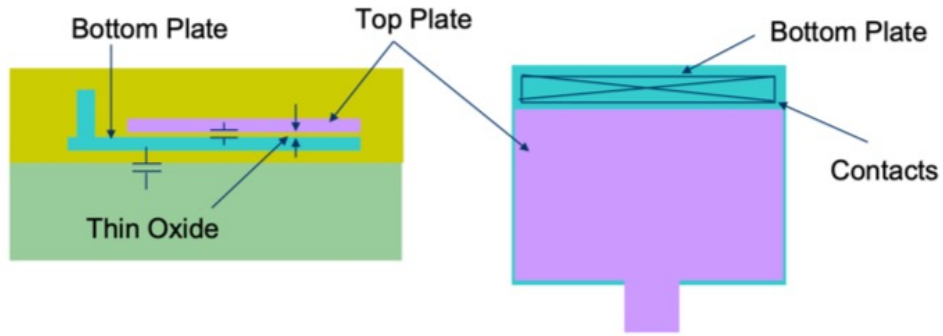
# Capacitor

## Basics

# Capacitor

## Implementation

**Model**



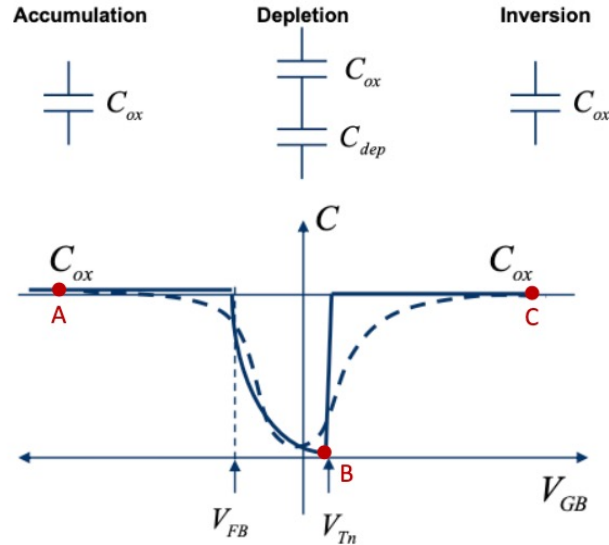
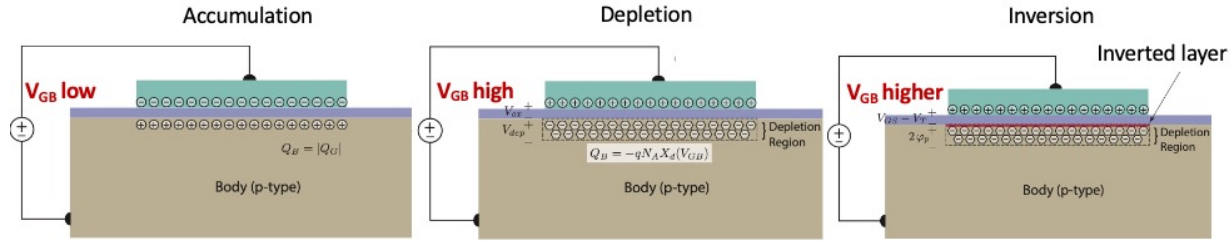
# Capacitor

## Summary of all options

Type	C [aF/ $\mu\text{m}^2$ ]	V <sub>C</sub> [ppm/V]	T <sub>C</sub> [ppm/ $^{\circ}\text{C}$ ]
Gate (thick oxide)	8000	Huge	Big
MIM (option)	4000		
MOM (> 5 metals)	~2000		
Poly-poly (option)	1000	10	25
Metal-metal	50	20	30
Metal-substrate	30		
Metal-poly	50		
Poly-substrate	120		
Junction capacitors	~ 1000	Big	Big

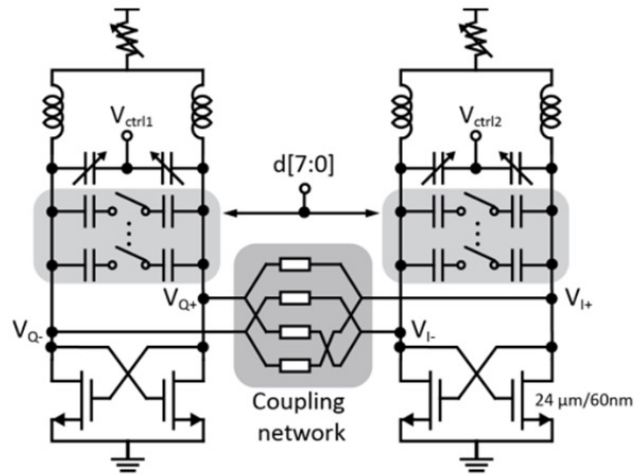
# Capacitor

## MOScap

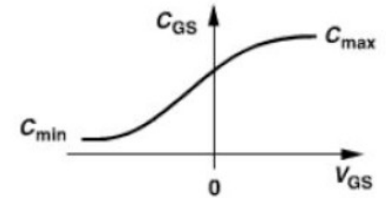
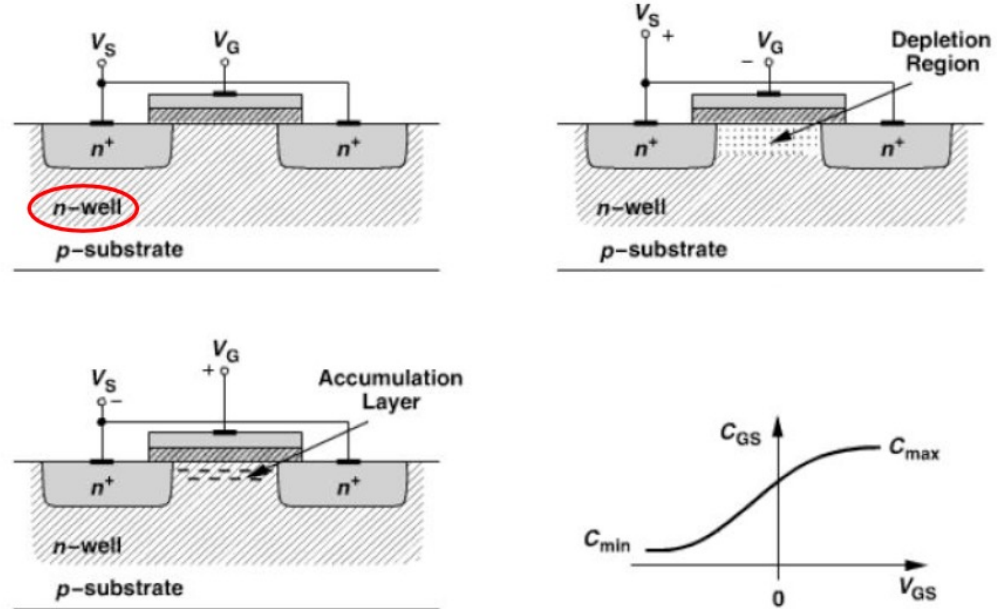


# Variable capacitor (or varactor)

## RF voltage-controlled oscillator (VCO)

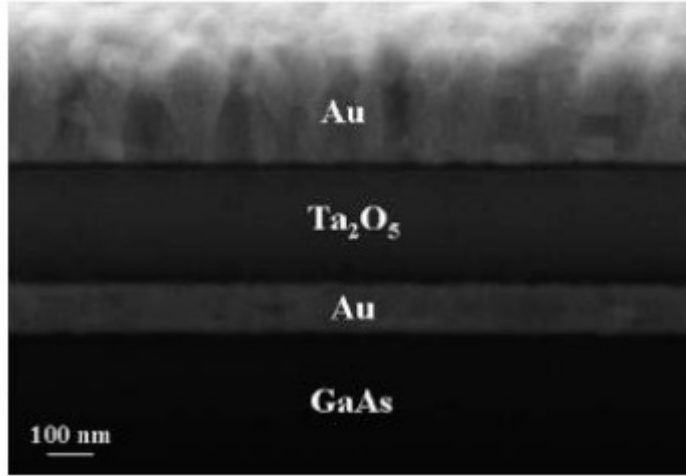


## Accumulation-mode MOS (A-MOS, or NMOS-in-Nwell)

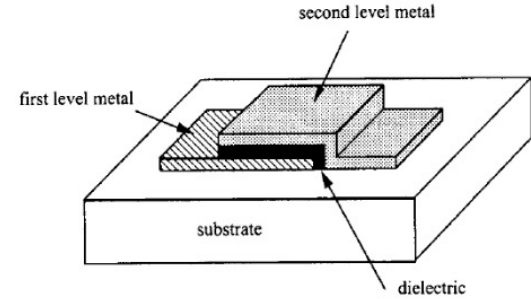


# MIMCap = Metal-insulator-metal capacitor

Not all processes have such an option...



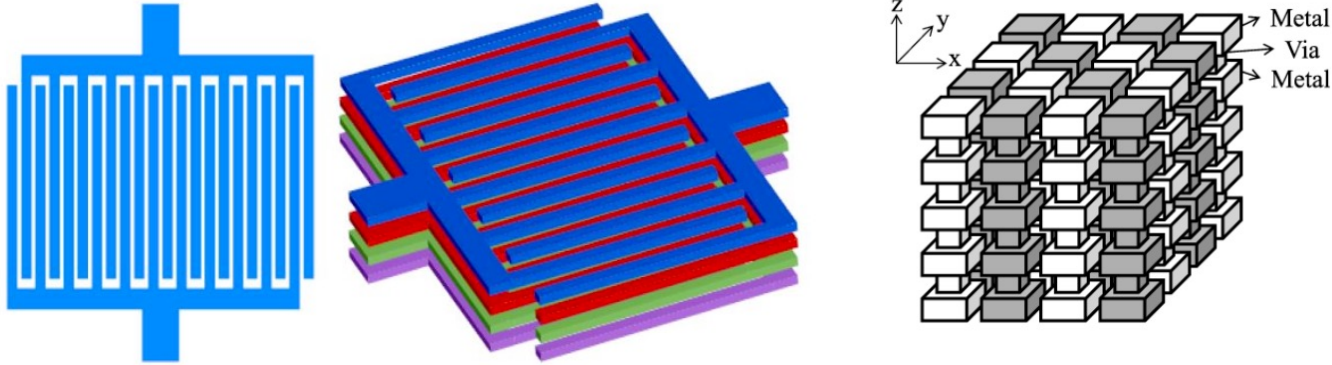
25 nm thin dielectric



Offered in CMOS 180nm at  $1\sim 2 \text{ fF}/\mu\text{m}^2$  but not the others per experience.

# MOMcap = Metal-oxide-metal capacitor

Come for free!



- Capacitance density:  $1\sim 3\text{ fF}/\mu\text{m}^2$ , lithography dependent.
- Watch out for the breakdown voltage.
- “CRTMOM” is pretty popular (interleaved with  $90^\circ$  rotation); can do custom post-layout extraction (pretty accurate compared to EM-simulation)
- Watch out for the parasitics, CMOS 90nm: 5%, CMOS 28nm: 1% (1% is a really good number!)
- Can bundle with MOScap and MIMcap as a high-density cell for supply decoupling unit.



# Capacitor layout

Always employ unit cell and translational matching

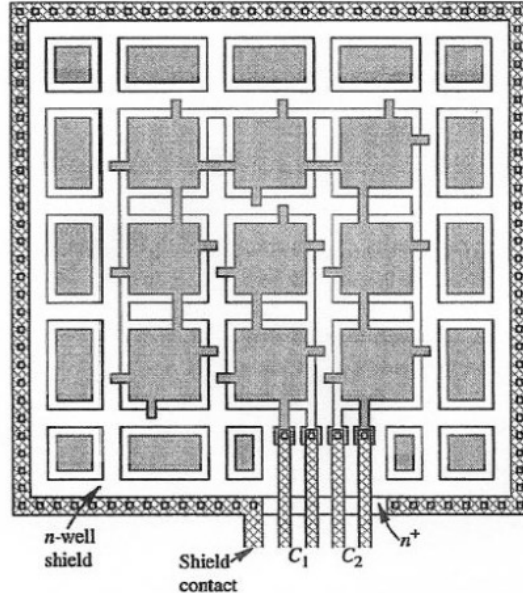


Figure 6.17 A layout of two capacitors with ratio  $C_2/C_1 = 1/2$ , incorporating several of the techniques discussed in the text (adapted from Ref. 23).

- Unit elements
- Shields:
  - Etching
  - Fringing fields
- “Common-centroid”
- Wiring and interconnect parasitics

Ref.: Y. Tsividis, “Mixed Analog-Digital VLSI Design and Technology,” McGraw-Hill, 1996.

# Capacitor Model

- A 3-terminal distributed model.
- $R_s$  represents series resistance of metal layers comprising the MoM structure.
- $C_{sub1,2}$  and  $R_{sub1,2}$  represent bottom plate caps to substrate and loss.
- $C_s$  is MoM cap value including fringing.
- Include series inductor to model phase shifts at very high-frequency ( $> 30$  GHz).

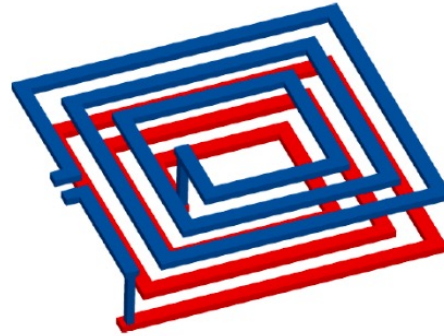
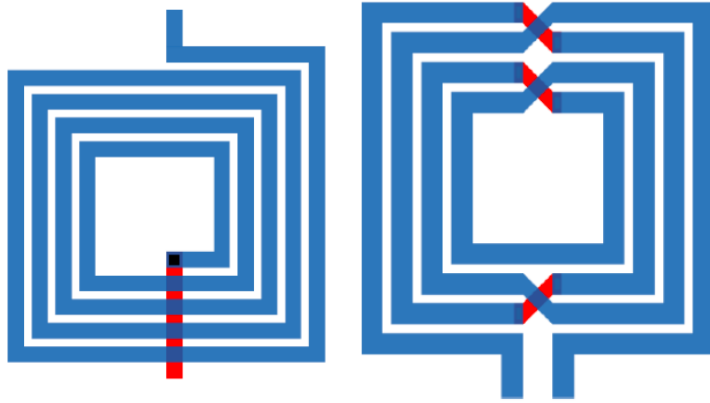
# Inductors

# Inductors

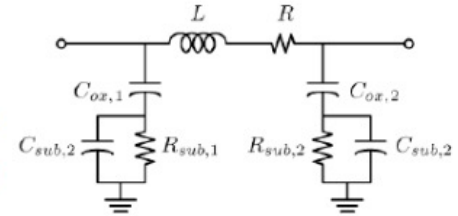
## Basics

# Inductors

## Layout

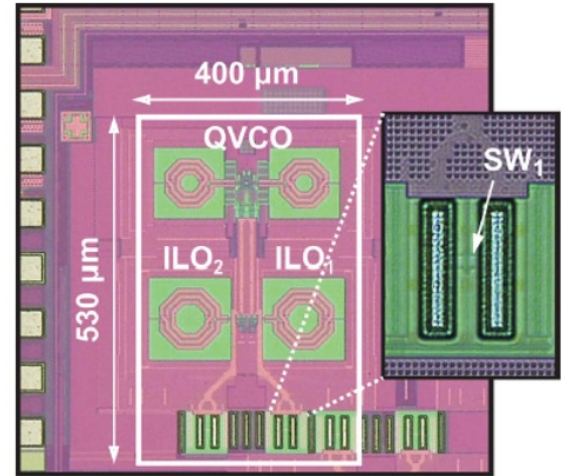
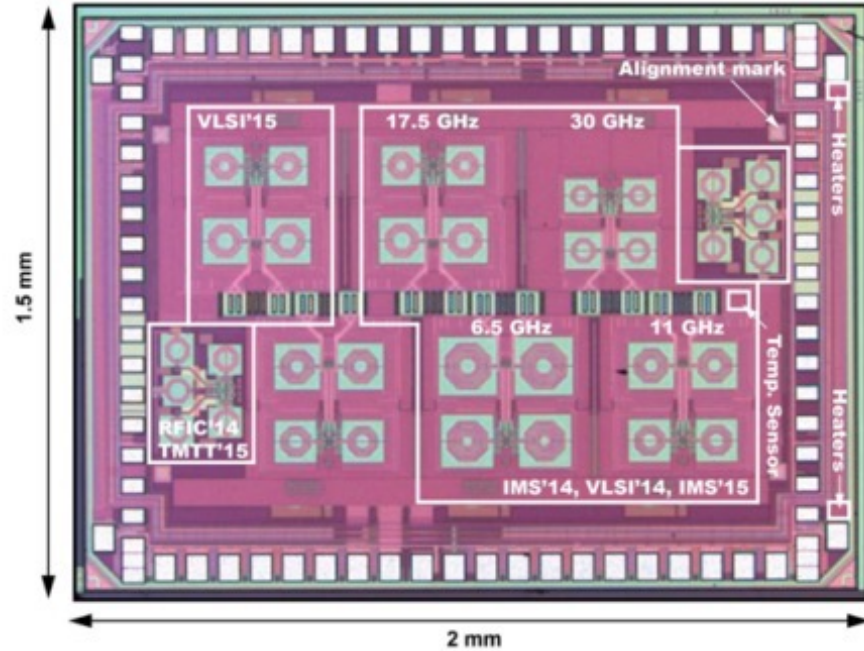


Lumped model

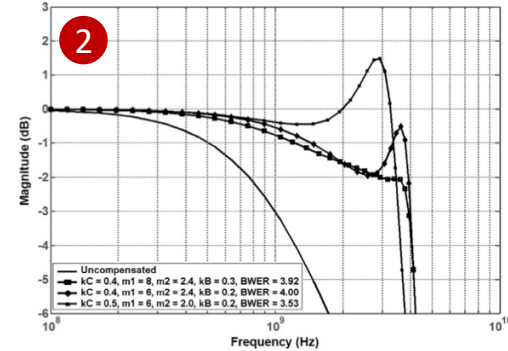
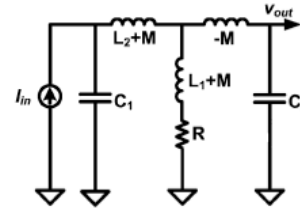
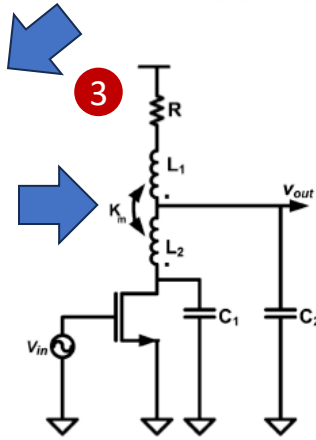
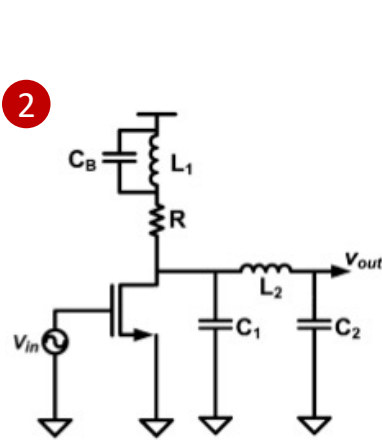
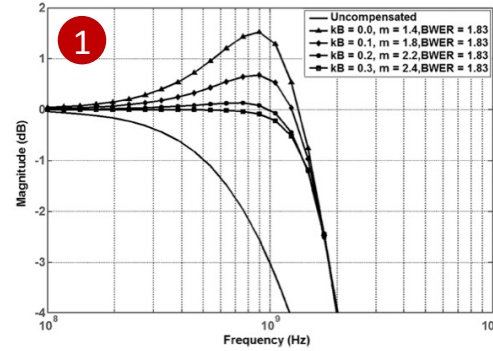
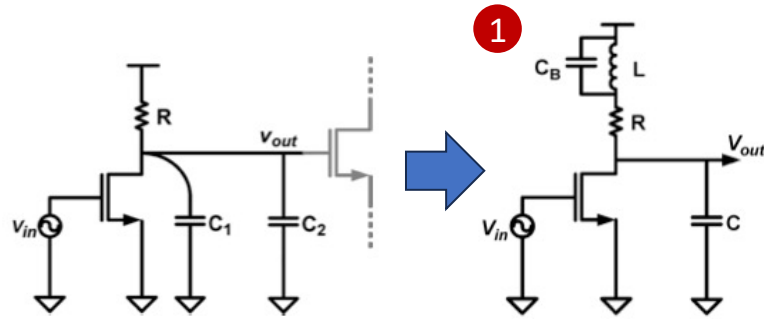


- Widely used in RF designs,  $L$  ranges from 30 pH  $\sim$  10 nH depending on the frequency, e.g. 500 MHz  $\sim$  300 GHz
- Implemented with top metal layers
  - Usually thick metals for power distribution, can be as thick as 4  $\mu\text{m}$
  - Away from substrate to avoid high parasitics capacitance
  - Si substrate must be “undoped” one (resistivity = 10 ohm-cm) to minimize losses
  - Requires E&M simulation (solving Maxwell Equations); usually the inductance is very accurate but  $Q$  requires iterations

# Example

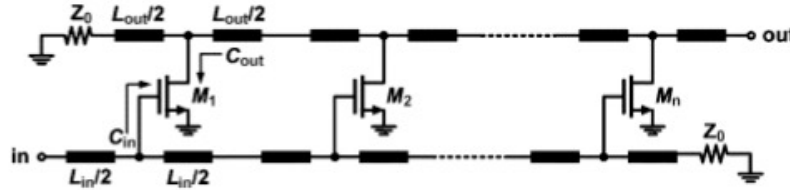


# Shunt peaking for bandwidth extension

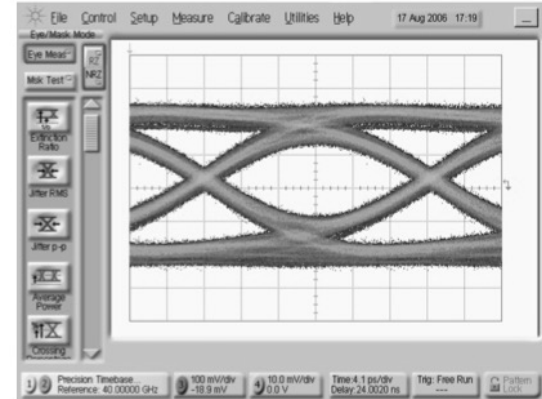
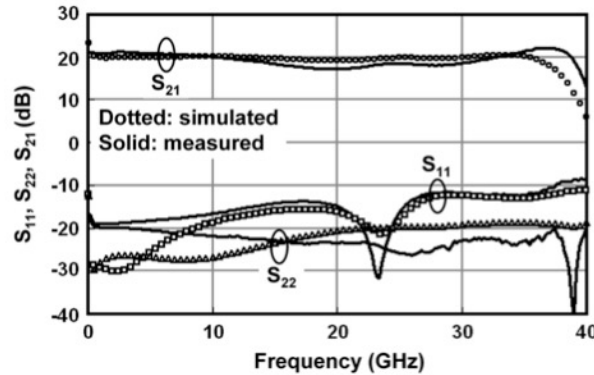
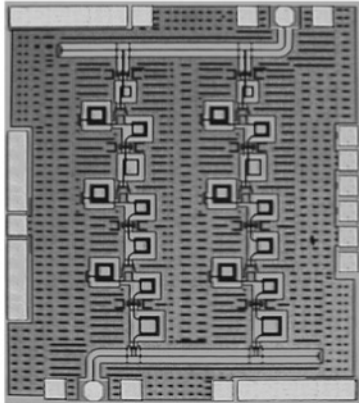


# Further distribution of capacitance

## Distributed amplifier



**20 dB, 39.4 GHz in 180-nm CMOS with  $f_T = 50$  GHz**



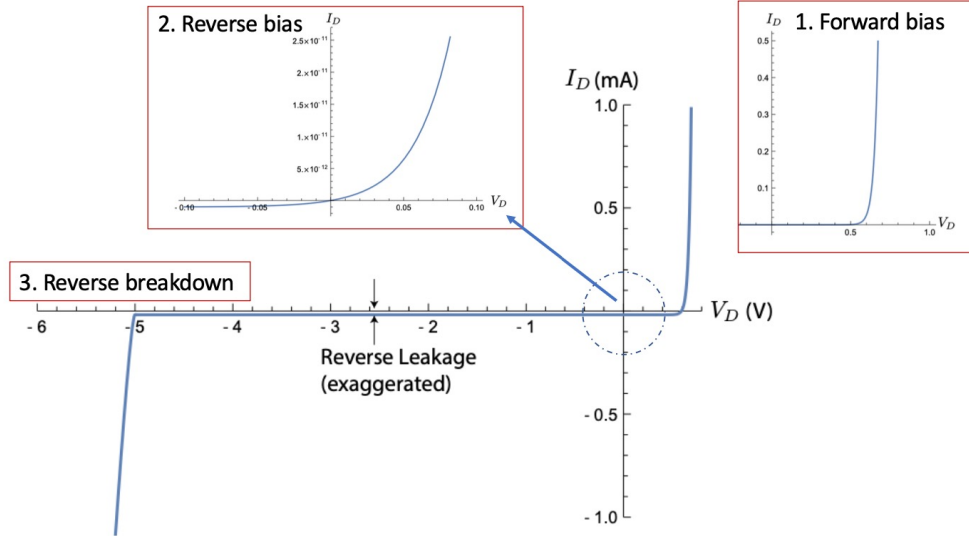
Chien, ISSCC'07



# Diodes

# Diodes

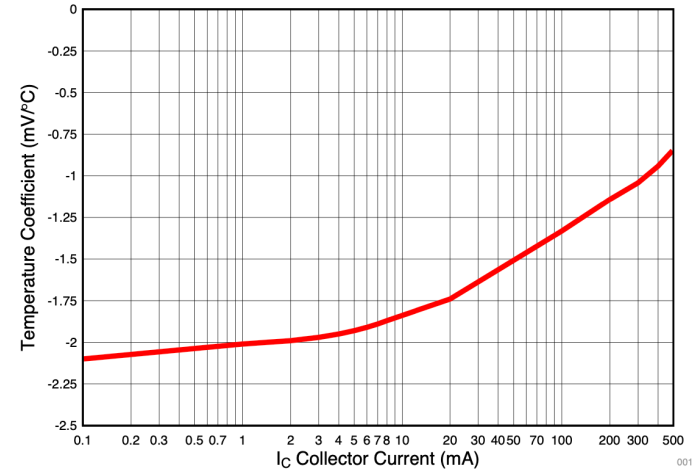
## I-V characteristics



$$I_D = I_S \left( e^{\frac{V_D}{V_T}} - 1 \right),$$

$$V_D = \frac{kT}{q} \ln \left( \frac{I}{I_S} \right)$$

$$\frac{\partial I_S}{\partial T} = 0.1 \text{ pA/K}$$



Source: TI App Note

# Diodes of various types

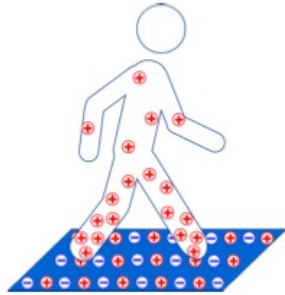
## Cross-section view and layout

P+/Nwell

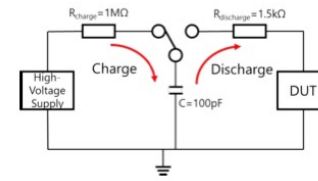
N-well/PSUB

N+/PSUB

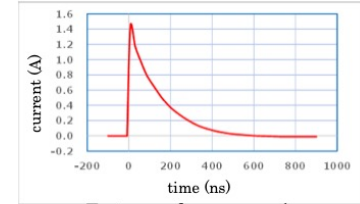
# ESD protection



## ESD testing standards

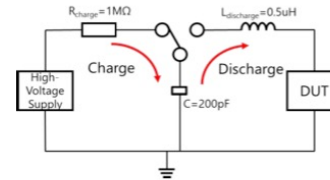


Test circuit example

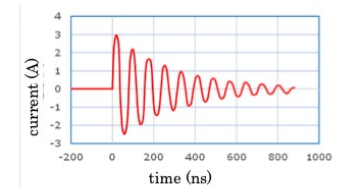


Test waveform example

Figure 2.3 Human body model (HBM)



Test circuit example



Test waveform example

Figure 2.4 Machine model (MM)

<https://toshiba.semicon-storage.com>

- ESD by itself is a dedicated research topic!
- In the company, there will be a dedicated ESD team for the management and testing.
- There are specific I/O pads with ESD cells offered by the foundry.
- Do not ignore its parasitics capacitance!

## BJTs

# Both NPN and PNP possible

## Cross-section view and layout

Vertical PNP (vpnp)

Horizontal NPN (need triple-well process)

- $\beta \sim 10$ , relative low speed due to wide base. Mainly for bandgap reference.
- Mostly PNP in CMOS.